FAMILY OF MICROPOWER RAIL-TO-RAIL OUTPUT
OPERATIONAL AMPLIFIERS

FEATURES

- **BiMOS Rail-to-Rail Output**
- **Input Bias Current . . . 1 pA**
- **High Wide Bandwidth . . . 160 kHz**
- **High Slew Rate . . . 0.1 V/μs**
- **Supply Current . . . 7 μA (per channel)**
- **Input Noise Voltage . . . 89 nV/$\sqrt{Hz}$**
- **Supply Voltage Range . . . 2.7 V to 16 V**
- **Specified Temperature Range**
  - **–40°C to 125°C . . . Industrial Grade**
  - **0°C to 70°C . . . Commercial Grade**
- **Ultra-Small Packaging**
  - **5 Pin SOT-23 (TLV27L1)**

APPLICATIONS

- **Portable Medical**
- **Power Monitoring**
- **Low Power Security Detection Systems**
- **Smoke Detectors**

DESCRIPTION

The TLV27Lx single supply operational amplifiers provide rail-to-rail output capability. The TLV27Lx takes the minimum operating supply voltage down to 2.7 V over the extended industrial temperature range, while adding the rail-to-rail output swing feature. The TLV27Lx also provides 160-kHz bandwidth from only 7 μA. The maximum recommended supply voltage is 16 V, which allows the devices to be operated from ($\pm$8-V supplies down to $\pm$1.35 V) two rechargeable cells.

The rail-to-rail outputs make the TLV27Lx good upgrades for the TLC27Lx family—offering more bandwidth at a lower quiescent current. The TLV27Lx offset voltage is equal to that of the TLC27LxA variant. Their cost effectiveness makes them a good alternative to the TLC/V225x, where offset and noise are not of premium importance.

The TLV27L1/2 are available in the commercial temperature range to enable easy migration from the equivalent TLC27Lx. The TLV27L1 is not available with the power saving/performance boosting programmable pin 8.

The TLV27L1 is available in the small SOT-23 package —something the TLC27(L)1 was not—enabling performance boosting in a smaller package. The TLV27L2 is available in the 3mm x 5mm MSOP, providing PCB area savings over the 8-pin SOIC and 8-pin TSSOP.

SELECTION GUIDE

DEVICE	V_S [V]	I_Q/ch [μA]	V_{ICR} [V]	V_{IO} [mV]	I_{IB} [pA]	GBW [MHz]	SLEW RATE [V/μs]	V_n, 1 kHz [nV/$\sqrt{Hz}$]
TLV27Lx	2.7 to 16	11	–0.2 to V_S+1.2	5	60	0.18	0.06	89
TLV238x	2.7 to 16	10	–0.2 to V_S–0.2	4.5	60	0.18	0.06	90
TLC27Lx	4 to 16	17	–0.2 to V_S–1.5	10/5/2	60	0.085	0.03	68
OPAx349	1.8 to 5.5	2	–0.2 to V_S+0.2	10	10	0.070	0.02	300
OPAx347	2.3 to 5.5	34	–0.2 to V_S+0.2	6	10	0.35	0.01	60
TLC225x	2.7 to 16	62.5	0 to V_S–1.5	1.5/0.85	60	0.200	0.02	19

NOTE: All dc specs are maximums while ac specs are typicals.

TLV27L1
TLV27L2

PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE CODE	SYMBOL	SPECIFIED TEMPERATURE RANGE	ORDER NUMBER	TRANSPORT MEDIA
TLV27L1CD	SOIC-8	D	27V1C	0°C to 70°C	TLV27L1CD	Tube
					TLV27L1CDR	Tape and Reel
TLV27L1CDBV	SOT-23	DBV	VBIC		TLV27L1CDBVR	Tape and Reel
					TLV27L1CDBVT	
TLV27L1ID	SOIC-8	D	27V1I	−40°C to 125°C	TLV27L1ID	Tube
					TLV27L1IDR	Tape and Reel
TLV27L1IDBV	SOT-23	DBV	VBII		TLV27L1IDBVR	Tape and Reel
					TLV27L1IDBVT	
TLV27L2CD	SOIC-8	D	27V2C	0°C to 70°C	TLV27L2CD	Tube
					TLV27L2CDR	Tape and Reel
TLV27L2ID	SOIC-8	D	27V2I	−40°C to 125°C	TLV27L2ID	Tube
					TLV27L2IDR	Tape and Reel

absolute maximum ratings over operating free-air temperature (unless otherwise noted)[†]

Supply voltage, V_S	16.5 V
Input voltage, V_I (see Note 1)	V_S
Output current, I_O	100 mA
Differential input voltage, V_{ID}	V_S
Continuous total power dissipation	See Dissipation Rating Table
Maximum junction temperature, T_J	150°C
Operating free-air temperature range, T_A: C suffix	0°C to 70°C
I suffix	−40°C to 125°C
Storage temperature range, T_{stg}	−65°C to 125°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	300°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: Relative to GND pin.

DISSIPATION RATING TABLE

PACKAGE	θ_{JC} (°C/W)	θ_{JA} (°C/W)	$T_A \leq 25°C$ POWER RATING	$T_A = 85°C$ POWER RATING
D (8)	38.3	176	710 mW	370 mW
DBV (5)	55	324.1	385 mW	201 mW
DBV (6)	55	294.3	425 mW	221 mW

recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, (V_S)	Dual supply	±1.35	±8	V
	Single supply	2.7	16	
Input common-mode voltage range		−0.2	V_S−1.2	V
Operating free-air temperature, T_A	C-suffix	0	70	°C
	I-suffix	−40	125	

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electrical characteristics at recommended operating conditions, V_S = 2.7 V, 5 V, and 10 V (unless otherwise noted)

dc performance

	PARAMETER	TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNIT
V_{IO}	Input offset voltage	V_{IC} = V_S/2, V_O = V_S/2, R_L = 100 kΩ, R_S = 50 Ω		25°C		0.5	5	mV
				Full range			7	
α_{VIO}	Offset voltage drift			25°C		1.1		µV/°C
CMRR	Common-mode rejection ratio	V_{IC} = 0 V to V_S–1.2 V, R_S = 50 Ω		25°C	71	86		dB
				Full range	70			
A_{VD}	Large-signal differential voltage amplification	$V_{O(PP)}$=V_S/2, R_L = 100 kΩ	V_S = 2.7 V, 5 V	25°C	80	100		dB
				Full range	77			
			V_S = ±5 V	25°C	77	82		
				Full range	74			

† Full range is –40°C to 125°C for I suffix.

input characteristics

	PARAMETER	TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
I_{IO}	Input offset current	V_{IC} = V_S/2, V_O = V_S/2, R_L = 100 kΩ, R_S = 50 Ω		≤25°C		1	60	pA
				≤70°C			100	
				≤125°C			1000	
I_{IB}	Input bias current			≤25°C		1	60	pA
				≤70°C			200	
				≤125°C			1000	
$r_{i(d)}$	Differential input resistance			25°C		1000		GΩ
C_{IC}	Common-mode input capacitance	f = 1 kHz		25°C		8		pF

power supply

	PARAMETER	TEST CONDITIONS		T_A†	MIN	TYP	MAX	UNIT
I_Q	Quiescent current (per channel)	V_O = V_S/2		25°C		7	11	µA
				Full range			16	
PSRR	Power supply rejection ratio (ΔV_S/ΔV_{IO})	V_S = 2.7 V to 16 V, No load, V_{IC} = V_S/2 V		25°C	74	82		dB
				Full range	70			

† Full range is –40°C to 125°C for I suffix.

TLV27L1
TLV27L2

electrical characteristics at recommended operating conditions, V_S = 2.7 V, 5 V, and $\pm$5 V (unless otherwise noted) (continued)

output characteristics

PARAMETER		TEST CONDITIONS		T_A[†]	MIN	TYP	MAX	UNIT
V_O	Output voltage swing from rail	V_{IC} = V_S/2, I_{OL} = 100 µA	V_S = 2.7 V	25°C		160	200	mV
				Full range			220	
			V_S = 5 V	25°C		85	120	
				Full range			200	
			V_S = $\pm$5 V	25°C		50	120	
				Full range			150	
		V_{IC} = V_S/2, I_{OL} = 500 µA	V_S = 5 V	25°C		420	800	
				Full range			900	
			V_S = $\pm$5 V	25°C		200	400	
				Full range			500	
I_O	Output current	V_O = 0.5 V from rail	V_S = 2.7 V	25°C		400		µA

† Full range is –40°C to 125°C for I suffix.

dynamic performance

PARAMETER		TEST CONDITIONS		T_A	MIN	TYP	MAX	UNIT
GBP	Gain bandwidth product	R_L = 100 kΩ, C_L = 10 pF, f = 1 kHz		25°C		160		kHz
SR	Slew rate at unity gain	$V_{O(pp)}$ = 1 V, R_L = 100 kΩ, C_L = 50 pF		25°C		0.06		V/µs
				–40°C		0.05		
				125°C		0.8		
ϕ_M	Phase margin	R_L = 100 kΩ, C_L = 50 pF		25°C		62		°
t_s	Settling time (0.1%)	$V_{(STEP)pp}$ = 1 V, A_V = –1, C_L = 50 pF, R_L = 100 kΩ	Rise	25°C		62		µs
			Fall			44		

noise/distortion performance

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
V_n	Equivalent input noise voltage	f = 1 kHz	25°C		89		nV/$\sqrt{Hz}$
I_n	Equivalent input noise current	f = 1 kHz	25°C		0.6		fA/$\sqrt{Hz}$

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TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	vs Common-mode input voltage	1, 2, 3
I_{IB}/I_{IO}	Input bias and offset current	vs Free-air temperature	4
V_{OH}	High-level output voltage	vs High-level output current	5, 7, 9
V_{OL}	Low-level output voltage	vs Low-level output current	6, 8, 10
I_Q	Quiescent current	vs Supply voltage	11
		vs Free-air temperature	12
	Supply voltage and supply current ramp up		13
A_{VD}	Differential voltage gain and phase shift	vs Frequency	14
GBP	Gain-bandwidth product	vs Free-air temperature	15
ϕ_m	Phase margin	vs Load capacitance	16
CMRR	Common-mode rejection ratio	vs Frequency	17
PSRR	Power supply rejection ratio	vs Frequency	18
	Input referred noise voltage	vs Frequency	19
SR	Slew rate	vs Free-air temperature	20
$V_{O(PP)}$	Peak-to-peak output voltage	vs Frequency	21
	Inverting small-signal response		22
	Inverting large-signal response		23
	Crosstalk	vs Frequency	24

INPUT OFFSET VOLTAGE
vs
COMMON-MODE INPUT VOLTAGE

V_S = 2.7 V
T_A = 25°C

Figure 1

INPUT OFFSET VOLTAGE
vs
COMMON-MODE INPUT VOLTAGE

V_S = 2.7 V
T_A = 25°C

Figure 2

INPUT OFFSET VOLTAGE
vs
COMMON-MODE INPUT VOLTAGE

V_S = ±5 Vdc
T_A = 25°C

Figure 3

TLV27L1
TLV27L2

TYPICAL CHARACTERISTICS

**INPUT BIAS AND INPUT
OFFSET CURRENT
vs
FREE-AIR TEMPERATURE**

Figure 4

**HIGH-LEVEL OUTPUT VOLTAGE
vs
HIGH-LEVEL OUTPUT CURRENT**

Figure 5

**LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT**

Figure 6

**HIGH-LEVEL OUTPUT VOLTAGE
vs
HIGH-LEVEL OUTPUT CURRENT**

Figure 7

**LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT**

Figure 8

**HIGH-LEVEL OUTPUT VOLTAGE
vs
HIGH-LEVEL OUTPUT CURRENT**

Figure 9

**LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT**

Figure 10

**QUIESCENT CURRENT
vs
SUPPLY VOLTAGE**

Figure 11

**QUIESCENT CURRENT
vs
FREE-AIR TEMPERATURE**

Figure 12

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TYPICAL CHARACTERISTICS

**SUPPLY VOLTAGE AND
SUPPLY CURRENT RAMP UP**

V_S = 0 to 15 V,
R_L = 100 Ω,
C_L = 10 pF,
T_A = 25°C

Figure 13

**DIFFERENTIAL VOLTAGE GAIN
AND PHASE SHIFT
vs
FREQUENCY**

V_S = 5 V
R_L = 100 kΩ
C_L = 10 pF
T_A = 25°C

Figure 14

**GAIN-BANDWIDTH PRODUCT
vs
FREE-AIR TEMPERATURE**

V_S = 2.7 V,
V_S = 5 V

Figure 15

**PHASE MARGIN
vs
LOAD CAPACITANCE**

V_S = 5 V
R_L = 100 kΩ
T_A = 25°C

Figure 16

**COMMON-MODE REJECTION RATIO
vs
FREQUENCY**

V_S = 5 V
T_A = 25°C

Figure 17

**POWER SUPPLY REJECTION RATIO
vs
FREQUENCY**

V_S = ±2.5 V,
T_A = 25°C

Figure 18

**INPUT REFERRED NOISE VOLTAGE
vs
FREQUENCY**

V_S = 5 V,
G = 2,
R_F = 100 kΩ

Figure 19

**SLEW RATE
vs
FREE-AIR TEMPERATURE**

SR+

SR−

V_S = 5 V
Gain = 1
V_O = 1
R_L = 100 kΩ
C_L = 50 pF

Figure 20

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TLV27L1
TLV27L2

TYPICAL CHARACTERISTICS

PEAK-TO-PEAK OUTPUT VOLTAGE
vs
FREQUENCY

$V_S = 15\ V$

$R_L = 100\ k\Omega,$
$C_L = 10\ pF,$
$THD+N <= 5\%$

$V_S = 5\ V$

$V_S = 2.7\ V$

V_{OPP} – Output Voltage Peak-to-Peak – V

f – Frequency – Hz

Figure 21

INVERTING SMALL-SIGNAL
RESPONSE

$V_I = 3\ V_{PP}$

Gain = –1,
$R_L = 100\ k\Omega,$
$C_L = 10\ pF,$
$V_S = 5\ V,$
$V_O = 3\ V_{PP},$
$f = 1\ kHz$

$V_O = 3\ V_{PP}$

Amplitude – V_{PP}

t – Time – μs

Figure 22

INVERTING LARGE-SIGNAL
RESPONSE

$V_I = 100\ mV_{PP}$

Gain = –1,
$R_L = 100\ k\Omega,$
$C_L = 10\ pF,$
$V_S = 5\ V,$
$V_O = 100\ mV_{PP},$
$f = 1\ kHz$

$V_O = 100\ mV_{PP}$

Amplitude – V_{PP}

t – Time – μs

Figure 23

CROSSTALK
vs
FREQUENCY

$V_S = 5\ V$
$R_L = 2\ k\Omega$
$C_L = 10\ pF$
$T_A = 25°C$
Channel 1 to 2

Crosstalk – dB

f – Frequency – Hz

Figure 24

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APPLICATION INFORMATION

offset voltage

The output offset voltage (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:

$$V_{OO} = V_{IO}\left(1 + \left(\frac{R_F}{R_G}\right)\right) \pm I_{IB+} \, R_S \left(1 + \left(\frac{R_F}{R_G}\right)\right) \pm I_{IB-} \, R_F$$
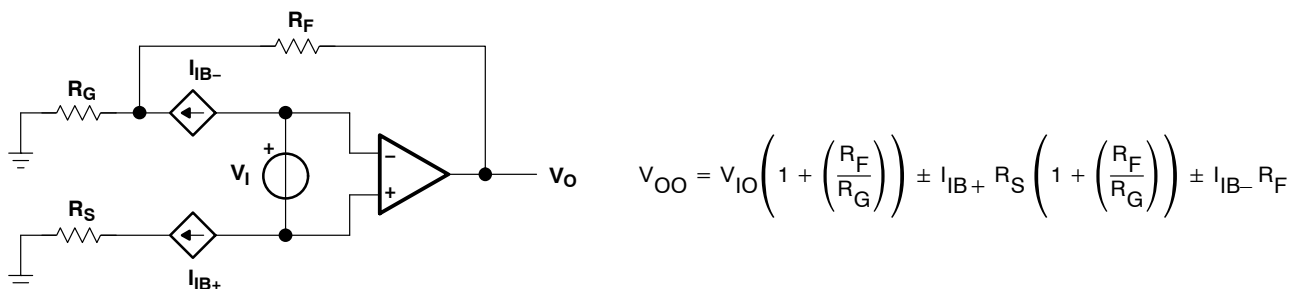
Figure 25. Output Offset Voltage Model

general configurations

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to accomplish this is to place an RC filter at the noninverting terminal of the amplifier (see Figure 26).
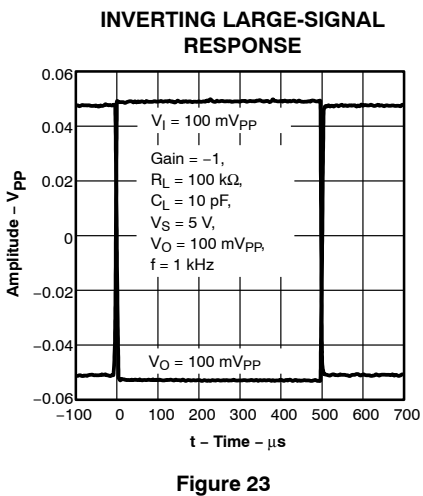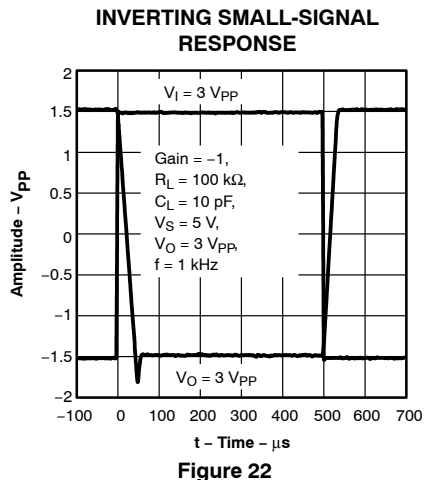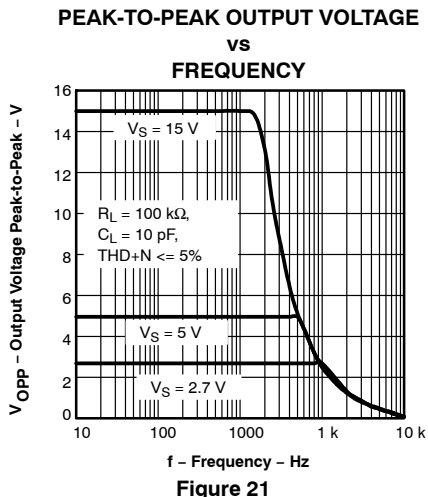

$$\frac{V_O}{V_I} = \left(1 + \frac{R_F}{R_G}\right)\left(\frac{1}{1 + sR1C1}\right)$$

$$f_{-3dB} = \frac{1}{2\pi R1C1}$$
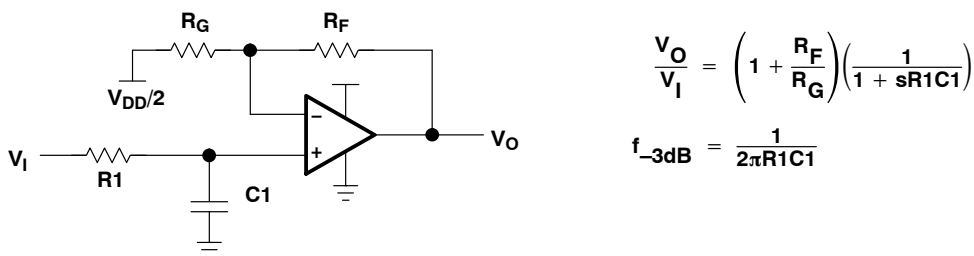
Figure 26. Single-Pole Low-Pass Filter

If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to do this can result in phase shift of the amplifier.

R1 = R2 = R
C1 = C2 = C
Q = Peaking Factor
(Butterworth Q = 0.707)

$$f_{-3dB} = \frac{1}{2\pi RC}$$

$$R_G = \frac{R_F}{\left(2 - \frac{1}{Q}\right)}$$

Figure 27. 2-Pole Low-Pass Sallen-Key Filter

APPLICATION INFORMATION

circuit layout considerations

To achieve the levels of high performance of the TLV27Lx, follow proper printed-circuit board design techniques. A general set of guidelines is given in the following.

- Ground planes—It is highly recommended that a ground plane be used on the board to provide all components with a low inductive ground connection. However, in the areas of the amplifier inputs and output, the ground plane can be removed to minimize the stray capacitance.

- Proper power supply decoupling—Use a 6.8-μF tantalum capacitor in parallel with a 0.1-μF ceramic capacitor on each supply terminal. It may be possible to share the tantalum among several amplifiers depending on the application, but a 0.1-μF ceramic capacitor should always be used on the supply terminal of every amplifier. In addition, the 0.1-μF capacitor should be placed as close as possible to the suppl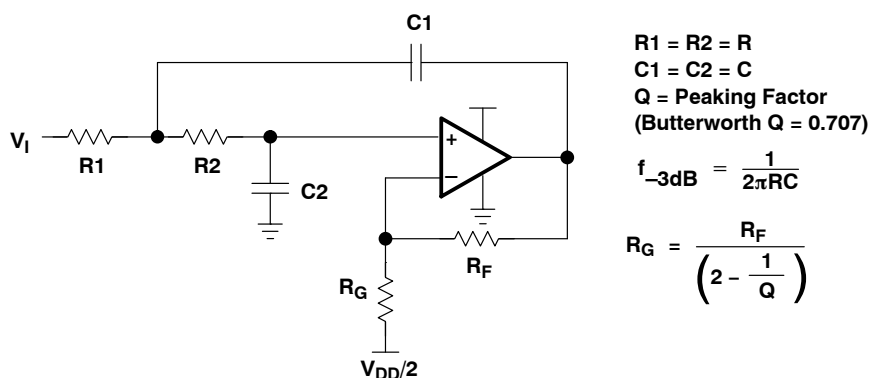y terminal. As this distance increases, the inductance in the connecting trace makes the capacitor less effective. The designer should strive for distances of less than 0.1 inches between the device power terminals and the ceramic capacitors.

- Sockets—Sockets can be used but are not recommended. The additional lead inductance in the socket pins will often lead to stability problems. Surface-mount packages soldered directly to the printed-circuit board is the best implementation.

- Short trace runs/compact part placements—Optimum high performance is achieved when stray series inductance has been minimized. To realize this, the circuit layout should be made as compact as possible, thereby minimizing the length of all trace runs. Particular attention should be paid to the inverting input of the amplifier. Its length should be kept as short as possible. This will help to minimize stray capacitance at the input of the amplifier.

- Surface-mount passive components—Using surface-mount passive components is recommended for high performance amplifier circuits for several reasons. First, because of the extremely low lead inductance of surface-mount components, the problem with stray series inductance is greatly reduced. Second, the small size of surface-mount components naturally leads to a more compact layout thereby minimizing both stray inductance and capacitance. If leaded components are used, it is recommended that the lead lengths be kept as short as possible.

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APPLICATION INFORMATION

general power dissipation considerations

For a given θ_{JA}, the maximum power dissipation is shown in Figure 28 and is calculated by the following formula:

$$P_D = \left(\frac{T_{MAX} - T_A}{\theta_{JA}}\right)$$

Where:

P_D = Maximum power dissipation of TLV27Lx IC (watts)
T_{MAX} = Absolute maximum junction temperature (150°C)
T_A = Free-ambient air temperature (°C)
θ_{JA} = $\theta_{JC} + \theta_{CA}$
    θ_{JC} = Thermal coefficient from junction to case
    θ_{CA} = Thermal coefficient from case to ambient air (°C/W)

MAXIMUM POWER DISSIPATION
vs
FREE-AIR TEMPERATURE
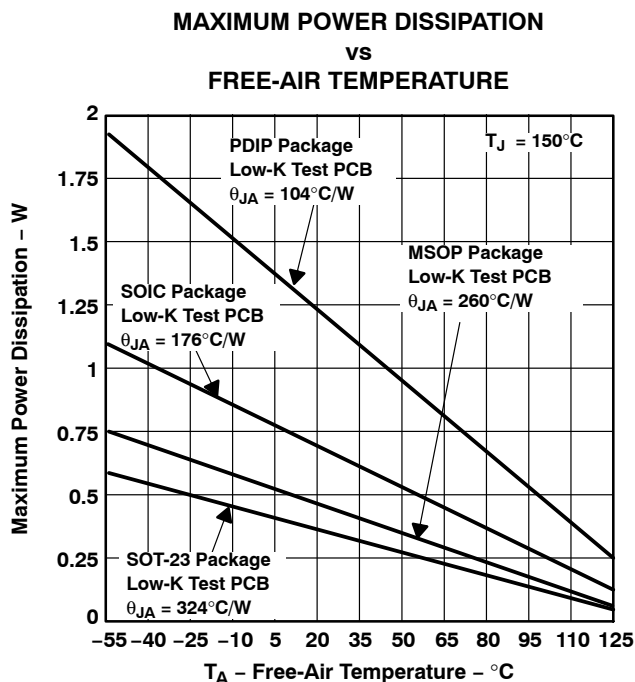
NOTE A: Results are with no air flow and using JEDEC Standard Low-K test PCB.

Figure 28. Maximum Power Dissipation vs Free-Air Temperature

NC – No internal connection
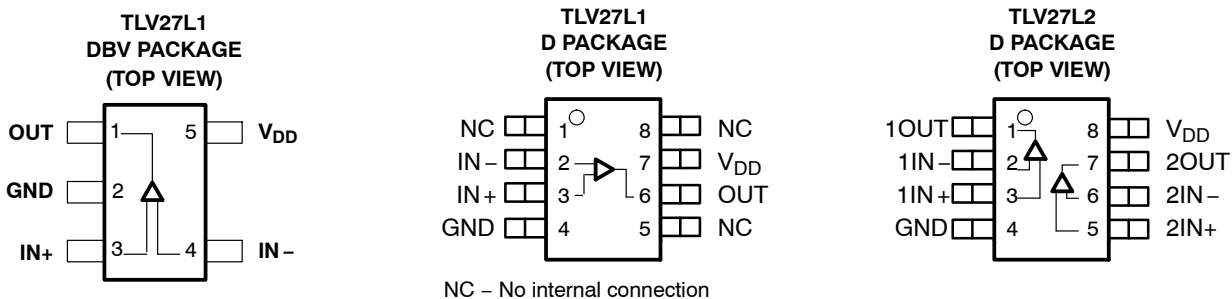
PACKAGING INFORMATION

Orderable Device	Status [1]	Package Type	Package Drawing	Pins	Package Qty	Eco Plan [2]	Lead/ Ball Finish	MSL Peak Temp [3]	Samples (Requires Login)
TLV27L1CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1CDBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1CDBVRG4	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1CDBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1CDBVTG4	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDBVRG4	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDBVTG4	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L1IDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2CDGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2CDGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2CDGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

Orderable Device	Status [1]	Package Type	Package Drawing	Pins	Package Qty	Eco Plan [2]	Lead/ Ball Finish	MSL Peak Temp [3]	Samples (Requires Login)
TLV27L2CDGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLV27L2IDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

[1] The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

[2] Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

[3] MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV27L1CDBVR	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TLV27L1CDBVT	SOT-23	DBV	5	250	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TLV27L1IDBVR	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TLV27L1IDBVT	SOT-23	DBV	5	250	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TLV27L1IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV27L2CDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLV27L2CDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLV27L2CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV27L2IDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLV27L2IDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TLV27L2IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV27L1CDBVR	SOT-23	DBV	5	3000	182.0	182.0	20.0
TLV27L1CDBVT	SOT-23	DBV	5	250	182.0	182.0	20.0
TLV27L1IDBVR	SOT-23	DBV	5	3000	182.0	182.0	20.0
TLV27L1IDBVT	SOT-23	DBV	5	250	182.0	182.0	20.0
TLV27L1IDR	SOIC	D	8	2500	340.5	338.1	20.6
TLV27L2CDGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TLV27L2CDGKR	VSSOP	DGK	8	2500	364.0	364.0	27.0
TLV27L2CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLV27L2IDGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TLV27L2IDGKR	VSSOP	DGK	8	2500	364.0	364.0	27.0
TLV27L2IDR	SOIC	D	8	2500	340.5	338.1	20.6

DBV (R−PDSO−G5)　　　　　PLASTIC SMALL−OUTLINE PACKAGE

4073253−4/K 03/2006

NOTES:　A.　All linear dimensions are in millimeters.
　　　　　B.　This drawing is subject to change without notice.
　　　　　C.　Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
　　　　　D.　Falls within JEDEC MO−178 Variation AA.

DGK (S−PDSO−G8)   PLASTIC SMALL−OUTLINE PACKAGE

0,65

8   5

0,38
0,25   ⊕ 0,13 Ⓜ

D 3,10   5,05
   2,90   4,75

1   4

C
3,10
2,90

0,23
0,13

Gauge Plane

0°−8°

0,25

0,70
0,40

1,10 MAX

0,15
0,05

Seating Plane

0,10

4073329/E 05/06

NOTES: A. All linear dimensions are in millimeters.
         B. This drawing is subject to change without notice.
         C▷ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall
             not exceed 0.15 per end.
         D▷ Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
         E. Falls within JEDEC MO−187 variation AA, except interlead flash.

D (R−PDSO−G8)   PLASTIC SMALL OUTLINE

0.197 (5,00)
0.189 (4,80)
C

8   5

0.244 (6,20)
0.228 (5,80)

0.157 (4,00)
0.150 (3,80)
D

Pin 1
Index Area

1   4

0.050 (1,27)

0.020 (0,51)
0.012 (0,31)

⊕ 0.010 (0,25) M

0.069 (1,75) Max

0.010 (0,25)
0.004 (0,10)

0.010 (0,25)
0.005 (0,13)

Gauge Plane

0.004 (0,10)

Seating Plane

0.010 (0,25)

0°−8°

0.050 (1,27)
0.016 (0,40)

4040047−3/M   06/11

NOTES: A. All linear dimensions are in inches (millimeters).
         B. This drawing is subject to change without notice.
         C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall
             not exceed 0.006 (0,15) each side.
         D. Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
         E. Reference JEDEC MS−012 variation AA.

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D (R−PDSO−G8)   PLASTIC SMALL OUTLINE

Example Board Layout
(Note C)

Stencil Openings
(Note D)

6x1,27

8x0,55

6x1,27

8x1,50

5,40

5,40

Example
Non Soldermask Defined Pad

Example
Pad Geometry
(See Note C)

0,60

1,55

Example
Solder Mask Opening
(See Note E)

0,07

All Around

4211283−2/E   08/12

NOTES: A.   All linear dimensions are in millimeters.
          B.   This drawing is subject to change without notice.
          C.   Publication IPC−7351 is recommended for alternate designs.
          D.   Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
               contact their board assembly site for stencil design recommendations. Refer to IPC−7525 for other stencil recommendations.
          E.   Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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